

Lecture

11

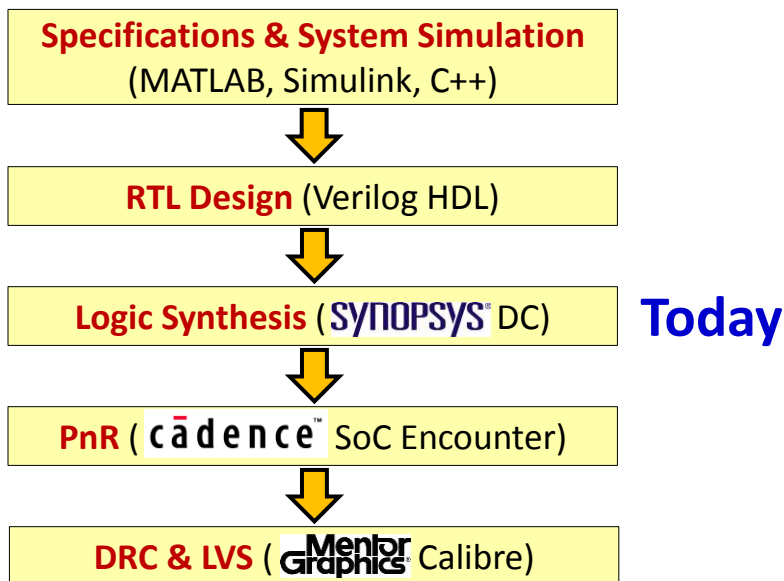
EEM216A
Fall 2012

Logic Synthesis

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VLSI Design Flow



11.2

Logic Synthesis

⇒ Conceptual overview & tool setup

- RTL modification
- Technology libraries
- Design environment & constraints
- Major synthesis commands
- Gate-level simulation

11.3

Synthesis:

WHY, HOW, WHAT

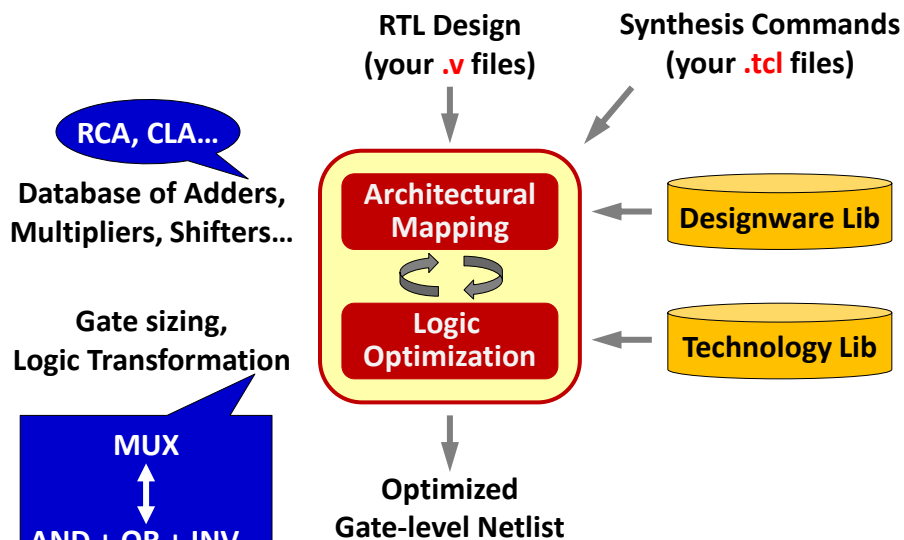
11.4

Why Synthesis?

- Quick (re)design time
- Separate functionality, technology-dependent parameters, and design constraints
- Fast timing/area/power estimates

11.5

How to do Synthesis?

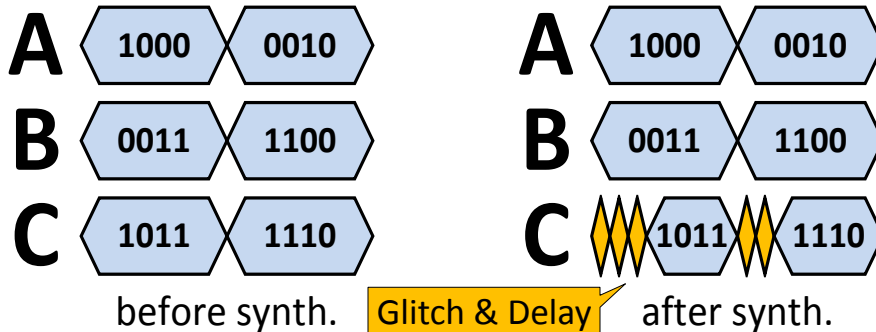


11.6

What can Synthesis Provide?

- Timing/Area/Power reports
- Gate-level netlist (**.vg**) & timing info. in standard delay format (**.sdf**) for timing-aware simulation

assign C = A + B ;



11.7

Script Example (.tcl Files)

```

remove_design -all
source ./SET_CON/T40GP_LibSetup.tcl
lappend search_path ./HDL ./SYN
analyze -format verilog Adder.v
set PROCESS_T40GP
set DESIGN_NAME ADD
elaborate $DESIGN_NAME
link
set_operating_conditions -min ff1p16vn40c -max ss0p95v125c
set_wire_load_selection_group "predcaps"
set TCLK 1.9
set TCU 0.1
source ./SET_CON/T40GP_VarSetup.tcl
set_false_path -from [get_ports ACLR_]
uniquify
compile -only_design_rule
compile_ultra -no_autoungroup
compile -inc -only_hold_time
set_fix_multiple_port_nets -all -buffer_constants [get_designs *]
remove_unconnected_ports -blast_buses [get_cells -hier]
report_timing -path full -delay min -max_paths 10 > $LOGPATH$TOPLEVEL$PROCESS.holdtiming
report_timing -path full -delay max -max_paths 10 > $LOGPATH$TOPLEVEL$PROCESS.setuptiming
report_area -hierarchy > $LOGPATH$TOPLEVEL$PROCESS.area
report_power -hier -hier_level 2 > $LOGPATH$TOPLEVEL$PROCESS.power

```

1. Setup Technology Library

2. Read RTL design & do architectural mapping

3. Setup design constraint

4. Compile design

5. Get Reports

11.8

Tool Setup on Workstation

Set path = (/w/apps3/Synopsys/Design_Compiler/vF-2011.09/bin \$path)
Source /w/apps3/Synopsys/Design_Compiler/vF-2011.09/SETUP

```
*****  
* SEASnet Computing Access  
*****  
* Priority is given both on the server and in the student labs to those  
* students doing coursework. Computer time is not guaranteed for other  
* each department.  
*****  
* For assistance please contact help@seas.ucla.edu or call 206-6864.  
*****  
[ee216ota@eeapps02 ~]$ set path = (/w/apps3/Synopsys/Design_Compiler/vF-2011.09/bin $path)  
[ee216ota@eeapps02 ~]$ source /w/apps3/Synopsys/Design_Compiler/vF-2011.09/SETUP  
[ee216ota@eeapps02 ~]$ dc_shell  
Design Compiler Graphical  
DC Ultra (TM)  
DFTMAX (TM)  
Power Compiler (TM)  
DesignWare (R)  
DC Expert (TM)  
Design Vision (TM)  
HDL Compiler (TM)  
VHDL Compiler (TM)  
DFT Compiler  
Library Compiler (TM)  
Design Compiler (R)
```

Now you're ready to use [Synopsys Design Compiler](#)

11.9

Logic Synthesis

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⇒ RTL modification

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11.10

The Importance of HDL Coding Styles

Different coding styles that are functionally eqvl. may be mapped into HWs having different timing/area

$$E = A + B + C + D$$

$$E = (A + B) + (C + D)$$



Good coding style: **Better starting point for DC to reach optimal gate-level results**

11.11

Enable Cool Features from DC

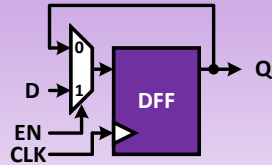
- #1: Gated Clock**
- #2: Synopsys Directives**

11.12

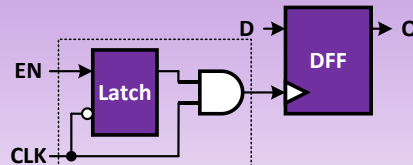
Go Low-Power using **Clock Gating**

30~50% Power Savings @ Block Level

STILL toggles when EN = 0



NO toggle when EN = 0



```
always@(posedge CLK)
  if(EN) Q <= D ;
  else  Q <= Q ;
```



```
always@(posedge CLK)
  if(EN) Q <= D ;
```

Two steps for gated-clock syn.: (1) Delete **Else** statements in seq. logic; (2) DC commands (shown later)

11.13

Case Directives Supported by DC (1)

tell DC to interpret incomplete case specially

Keyword: **// synopsys_parallel_case**

// synopsys_full_case

(RTL simulator ignores, but DC understands)

```
reg A, B;
wire [2:0] SEL ;
always@(*)
{A, B} = 2'b0 ;
casez(SEL) // synopsys_parallel_case
  3'b?11: B = 1 ;
  3'b1??: A = 1 ;
endcase
```

Put it after the
case declaration



11.14

Case Directives Supported by DC (2)

- **Synopsys_full_case**
 - Treats o/p of unstated case items as **don't care**
 - Ignored if the case statement is complete
 - As a result, **no latch** is created
- **Synopsys_parallel_case**
 - Treats all case items as **non-overlapped**; implement each of them separately
 - As a result, **no priority logic** is created

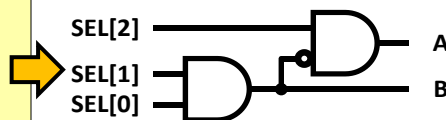
Note: Potential discrepancy b/w
RTL simulation and gate-level simulation

11.15

Example: RTL the Same, Gate-Level Different

```
reg A, B;  
wire [2:0] SEL ;  
always@(*)  
{A, B} = 2'b0 ;  
casez(SEL)  
  3'b?11: B = 1 ;  
  3'b1??: A = 1 ;  
endcase
```

Lec. 8.46



Priority Logic

```
reg A, B;  
wire [2:0] SEL ;  
always@(*) // synopsys_parallel_case  
{A, B} = 2'b0 ;  
casez(SEL)  
  3'b?11: B = 1 ;  
  3'b1??: A = 1 ;  
endcase
```



Independent Logic

11.16

Logic Synthesis

- Conceptual overview & tool setup
- RTL modification

⇒ Technology libraries

- Design environment & constraints
- Major synthesis commands
- Gate-level simulation

11.17

What are Technology Libraries About? (1)

```
library ("saed32rvt_ss0p95v125c") {  
  operating_conditions (ss0p95v125c) {  
    process : 0.99;  
    voltage : 0.950000;  
    temperature : 125.000000;  
    tree_type : "best_case_tree";  
  }  
  wire_load (ForQA) {  
    capacitance : 0.029396;  
    resistance : 2.273000e-03;  
    area : 0.010000;  
    slope : 30.285426;  
    fanout_length ("1", "8.2750360");  
    ...  
  }  
  wire_load ("8000") {  
    ...  
  }  
  ...  
  wire_load_selection (predcaps) {  
    wire_load_from_area (0.000000, 200.000000, "ForQA");  
    wire_load_from_area (200.000000, 8000.000000, "8000");  
    ...  
  }  
}
```

Operating condition
(process, voltage, temp.)

Wire-load model
(rough est. of wiring
parasitics before PnR)

Wire-load selection
(based on synth.
area of your design)

11.18

What are Technology Libraries About? (2)

detailed information of each standard cell

Cell Name
Area
Functionality
Pin Capacitance
Leakage Power
7x7 Look-up Tables (LUT) for Dynamic Power &
Propagation Delay of all i/o Paths
(idx1: input transition time; idx2: o/p capacitance)

11.19

What are Technology Libraries About? (3)

```
cell (AND2X1_RVT) {  
  area : 2.033152; cell_leakage_power : 1.404953e+05;  
  leakage_power () { when : "!A1&!A2"; value : "5.6919234e+04"; }  
  ...  
  pin (Y) {  
    related_power_pin : "VDD"; related_ground_pin : "VSS"; direction : "output";  
    ...  
    max_capacitance : 8.000000; max_transition : 0.194746;  
    internal_power () {  
      related_pin : "A1";  
      rise_power ("power_outputs_1") {  
        index_1 ("0.016, 0.032, 0.064, 0.128, 0.256, 0.512, 1.024");  
        index_2 ("0.1, 0.25, 0.5, 1, 2, 4, 8");  
        values ("0.9612064, 0.9624440, 0.9509065, 0.9587265, 0.9569466, 0.9523558, 0.9245481", \  
          ...  
        )  
      }  
      timing () {  
        related_pin : "A1";  
        cell_rise ("del_1_7_7") {  
          index_1 ("0.016, 0.032, 0.064, 0.128, 0.256, 0.512, 1.024");  
          index_2 ("0.1, 0.25, 0.5, 1, 2, 4, 8");  
          values ("0.0400985, 0.0415339, 0.0437422, 0.0477826, 0.0546696, 0.0667816, 0.0888995", \  
            ...  
          )  
        }  
      }  
    }  
  }  
}
```

11.20

Access Synopsys 32nm Technology Libs.

Path: /w/apps2/public.2/tech/synopsys/32-28nm/SAED32_EDK/lib/stdcell_**rvt**/db_nldm
(rvt: regular- V_{th} . Can change to hvt and lvt for high- V_{th} and low- V_{th} libraries)

There you can find various libraries, each with different design corners. We select **saed32rvt_ff1p16vn40c.db** for hold-time analysis & **saed32rvt_ss0p95v125c.db** for setup-time analysis

Note: .db files are only used by DC. If you want to read library information as in page 11.18~20, access the corresponding .lib files

11.21

Setup Technology Libraries

```
set search_path "$search_path \  
/w/apps2/public.2/tech/synopsys/32-28nm/SAED32_EDK/lib/stdcell_rvt/db_nldm"  
set target_library "saed32rvt_ff1p16vn40c.db saed32rvt_ss0p95v125c.db"  
set link_library "* saed32rvt_ff1p16vn40c.db saed32rvt_ss0p95v125c.db dw_foundation.sldb"  
set synthetic_library "dw_foundation.sldb"  
  
# Define work path (note: The work path must exist, so you need to create a folder WORK first)  
define_design_lib WORK -path ./WORK  
set alib_library_analysis_path "/alib-52/"
```

(save above scripts in a .tcl file, and source it after you open DC)

- DC follows **search_path** to find libraries you specify
- DC uses cells in target library for logic optimization (so we need to do **set_target_library** first)
- **Remember:** Use backslash \ before changing lines to avoid compilation errors.

11.22

Logic Synthesis

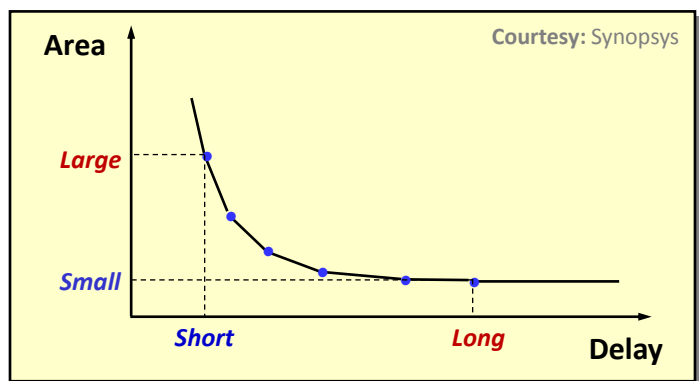
- Conceptual overview & tool setup
- RTL modification
- Technology libraries

⇒ Design environment & constraints

- Major synthesis commands
- Gate-level simulation

11.23

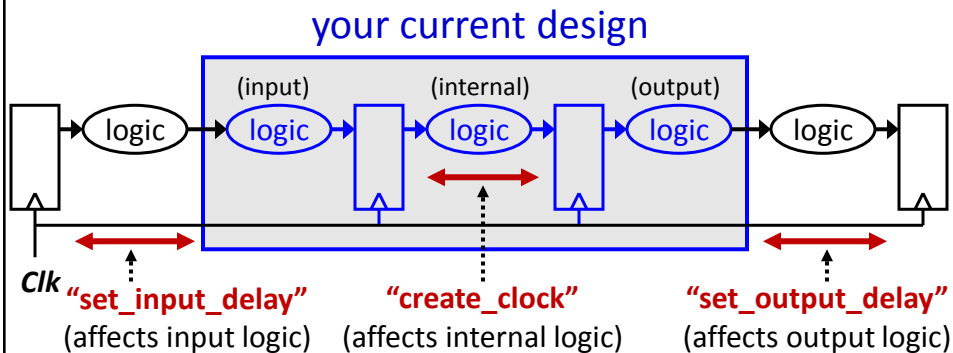
Synthesis is Constraint-Driven



You set the goals (through **design constraints**)
DC optimizes the design to meet your goals

11.24

Logic Synthesis is Timing-Driven



- **Input delay:** Arrival of an external path w.r.t. the clock edge
- **Output delay:** timing from an o/p (of **current design**) to a register i/p (of other submodules)

11.25

Describe Design Constraints

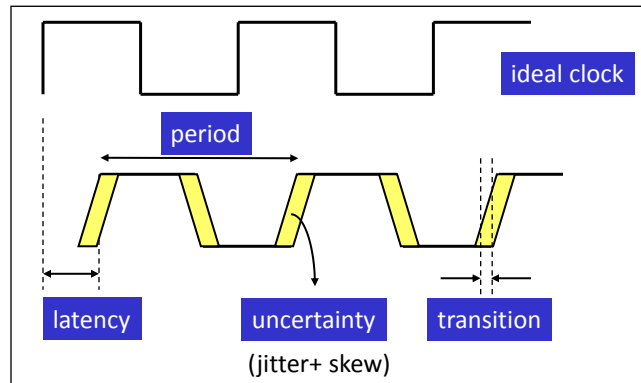
- **Clocks**
 - Period, latency, uncertainty
- **Design rules**
 - Maximum transition
 - Maximum capacitance
 - Maximum fanout
- **Input-related**
 - Driving cells, Input delay
- **Output-related**
 - Load, Output delay
- **Exception paths**
 - False paths, multi-cycle paths
- **Optimization goal**
 - Maximum area/power

Especially important for bottom-up design methodology

Accurate constraints (not too tight/loose): Good integ. results

11.26

Ideal vs. Real Clock



Courtesy: Synopsys

11.27

Clock Description

- **create_clock**: define clock's waveform (e.g. period)

```
create_clock -name "CLK" -period 2.0 [get_ports "CLK"]
```

- **set_fix_hold**: respect the hold time requirement of all clocked flip-flops

```
set_fix_hold CLK
```

- **set_dont_touch_network**: do not buffer clock network

```
set_dont_touch_network [get_clocks "CLK"]
```

- Specify uncertainty (skew + jitter) of clock network

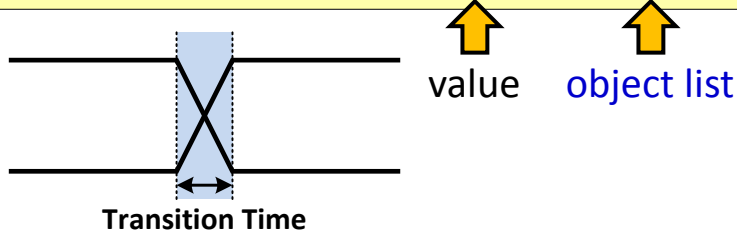
```
set_clock_uncertainty 0.1 [get_clocks "CLK"]
```

11.28

Maximum Transition

Sets the **max_transition** attribute to a specified value on specified clocks group, ports or designs.

```
set_max_transition 0.15 current_design
```



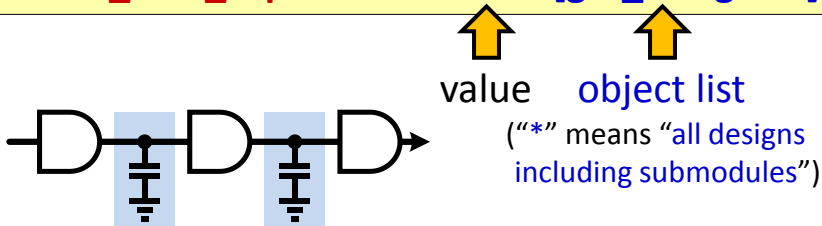
Tip: Check the **max_transition** variable in **.lib** file to setup more realistic numbers

8.29

Maximum Capacitance

Sets the **max_capacitance** attribute to a specified value on the specified input ports and designs.

```
set_max_capacitance 1.000 [get_designs *]
```



Tip: Check the **max_capacitance** variable in **.lib** file to setup more realistic numbers

8.30

Maximum Fanout

Sets the **max_fanout** attribute to a specified value on specified input ports and/or designs.

```
set_max_fanout 1.000 [get_designs *]
```

Total Fanout

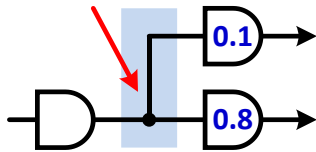
Load = 0.9

Gate Fanout

Load

value

object list



Tip: Check the **fanout_load** variable of cells in **.lib** file to setup more realistic numbers

8.31

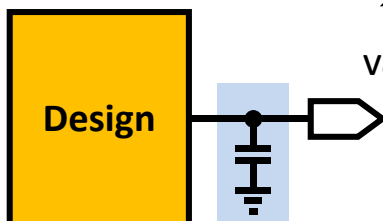
Output Load

Sets the **set_load** attribute to a specified value on specified output ports and/or designs.

```
set_load 1.000 [all_outputs]
```

value

object list



Tip: Check the **capacitance** variable of cell input pins in **.lib** file to setup more realistic numbers

8.32

Driving Cells

Sets attributes on input or inout ports of the current design, specifying that a library cell or output pin of a library cell drives specified ports.

```
set_driving_cell [-library lib] [-lib_cell  
lib_cell_name] [-pin pin_name]
```

```
set ALL_IN_BUT_CLK [  
remove_from_collection [all_inputs] "CLK"]  
  
set_driving_cell -no_design_rule -library  
saed32rvt_ss0p95v125c.db:saed32rvt_ss0p95v125c -  
lib_cell DFFASRX2_RVT -pin Q $ALL_IN_BUT_CLK
```

8.33

Input & Output Delay

Sets delay on pins or ports
relative to a (ideal) clock signal.

```
set_input_delay -max delay_value [-clock clock_name]  
set_input_delay -min delay_value [-clock clock_name]  
set_output_delay -max delay_value [-clock clock_name]  
set_output_delay -min delay_value [-clock clock_name]
```

```
set_input_delay 0.6 -clock "CLK" $ALL_IN_BUT_CLK  
set_input_delay -min 0.3 -clock "CLK" $ALL_IN_BUT_CLK  
set_output_delay 0.6 -clock "CLK" [all_outputs]  
set_output_delay -min 0.3 -clock "CLK" [all_outputs]
```

8.34

False Paths

Removes timing constraints from particular paths, but still needs to satisfy design rule (transition, capacitance, fanout).

set_false_path -from [from_list]

set_false_path -from [get_ports ACLR_]
 set_false_path -from [get_ports SI]

8.35

Maximum Area/Power

Optimization goals for your design (DC will do it best to satisfy them, w/o violating the three design rules)

set_max_area 0.0



desired area

set_max_total_power 0.0



desired power

11.36

Operating Condition

Defines the operating conditions for the current design.

```
set_operating_conditions  
[-min min_condition] [-max max_condition]
```

```
set_operating_conditions  
-min ff1p16vn40c -max ss0p95v125c
```



for hold-time check



for setup-time check

8.37

Wire-load Models and Modes

Specify a selection group to use for determining a wire load model to be assigned to designs and cells or to a specified cluster.

```
set_wire_load_selection_group group_name -max  
set_wire_load_selection_group group_name -min
```

```
set_wire_load_selection_group "predcaps"
```



DC will select proper model based on synthesis area (again, lookup this name in .lib file)

8.38

Logic Synthesis

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11.39

Preview: Use Synthesis Commands Properly

- Not all commands introduced are necessarily needed
- Things MUST do

Analyze, elaborate, and link design

Compile design

- Things can do based on your needs

Ungroup; Uniquify

Clock gating creation

Remove unconnected ports

Suggestion: Learn by playing w/ different commands and observing the differences

11.40

Analyze Designs

Analyzes the HDL files and stores the intermediate format in the specified library.

analyze [-format vhdl | verilog | sverilog] file_list

```
analyze -format verilog {adder.v}
```

8.41

Elaborate & Link Designs

Builds a design from the intermediate format of a Verilog module, a VHDL entity and architecture, or a VHDL configuration.

elaborate design_name

```
set DESIGN_NAME adder  
elaborate $DESIGN_NAME
```

8.42

Ungroup & Uniquify

Removes a level of hierarchy.

ungroup cell_list | -all [-flatten]

ungroup -flatten -all

Removes multiple-instantiated hierarchy in the current design by creating a unique design for each of the cell instances.

uniquify [-force] [-cell cell_list]

uniquify

8.43

Compile

Performs logic-level and gate-level synthesis and optimization on the current design.

compile [-no_design_rule | -only_design_rule | -only_hold_time] [-map_effort medium | high] [-boundary_optimization]

compile -only_design_rule
compile -map_medium high
compile -boundary_optimization
compile -only_hold_time

8.44

Gated Clock Creation

Four commands to construct clock gating
(note: Modify RTL as in pp. 13 to enable this feature)

```
set_clock_gating_registers -include_instances  
[all_registers -clock "CLK"]  
set_clock_gating_style -num_stages 2 -sequential_cell  
latch -minimum_bitwidth 8 -max_fanout 32  
insert_clock_gating -global  
propagate_constraints -gate_clock
```

11.45

Remove Unconnected Ports

Removes unconnected ports or pins from cells,
references, and subdesigns.

```
remove_unconnected_ports -blast_buses [get_cells -hier]
```

11.46

Report Designs

1. **report_timing** -path full -delay min -max_paths 10 -nworst 2 > **Design.holdtiming**
2. **report_timing** -path full -delay max -max_paths 10 -nworst 2 > **Design.setuptiming**
3. **report_area** -hierarchy > **Design.area**
4. **report_power** -hier -hier_level 2 > **Design.power**
5. **report_resources** > **Design.resources**
6. **report_constraint** -verbose > **Design.constraint**
7. **check_design** > **Design.check_design**
8. **check_timing** > **Design.check_timing**

11.47

Logic Synthesis

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➡ **Gate-level simulation**

11.48

Slight Modification of Original Testbench.v

- RTL simulation: **design.v** + **testbench.v**
- Gate-level simulation: **design.vg** + **design.sdf** (from DC) + **testbench.v** (w/ slight modification)

```
`timescale 1ns/10ps
module Test ;
  ADD AddSim(...);
  ...
  ...
  ...
  ...
  ...
endmodule
```



```
`timescale 1ns/10ps
module Test ;
  ADD AddSim(...);
  initial begin
    $sdf_annotate("design.sdf",
                  AddSim);
  end
  ...
  ...
endmodule
```

11.49

Any Questions?

The “Man” is always there for you, 24/7 😊
(type **man <syntax you want to lookup>** in DC)

```
eeapps.seas.ucla.edu - PuTTY
dc_shell> man set_input_delay
2. Synopsys Commands          set_input_delay          Command Reference

NAME
    set_input_delay
        Sets input delay on pins or input ports relative to a clock sig-
        nal.

SYNTAX
    status set_input_delay
            delay_value
            [-reference pin pin_port_name]
            [-clock clock_name]
```

or check out various user manuals online



11.50

Synthesis Summary

- **Synthesis tool makes VLSI design a lot easier**
 - Easy to use: **.v** files + **.tcl** files
 - Easy to switch designs to any technology by changing associated libraries
 - Easy to have accurate area/timing/power estimate
 - Easy to match the best archit. with design constraints
- **But... not a substitute for thinking**
 - Mind your coding styles
 - Handmade optimization (e.g. using shift-and-add for constant multiplication) still needed

11.51